

Advances in Science, Technology and Engineering Systems Journal Vol. 4, No. 4, 128-132 (2019)

ASTESJ ISSN: 2415-6698

www.astesj.com

Special Issue on Advancement in Engineering and Computer Science

Experimental Evaluation of Transmission between Two XBee Modules Using Radio-over-Fiber Technique

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ARTICLEINFO	A B S T R A C T
Article history: Received: 05 April, 2019 Accepted: 04 July, 2019 Online: 17 July, 2019	An experimental transmission between two XBee modules using Radio over Fiber technique is demonstrated. Data issued from an XBee module coded on a wireless microwave carrier at 2.44GHz is transmitted through an optical link of 25.24 km. The optical transmission is based on an external modulation scheme over a dispersive channel. Frequency response of
Keywords: Microwave signals Radio over fiber Optical communications XBee	the external modulation scheme is determined by simulation and experimentally. In particular, the chromatic dispersion of the optical fiber associated to its length allow the generation of a pass-band filter whose bandwidth is used to code the RF signal. The recovered RF signal exhibits a SNR of 27dB. Obtained results allow proposing this RoF scheme as a good contender for data transmission.

1. Introduction

This paper is an extension of the work originally presented at the 15th International Conference on Electrical Engineering, Computing Science and Automatic Control [1]. Now, the main contribution of this extended work resides in the fact that we are describing the mathematical procedure that allows determining the frequency response of the electro-optical system used for the transmission. Besides, the theoretical behavior of the optical communication system is validated by simulations. This work is situated on the field of Radio-over-Fiber (RoF), a technique that allows joining radio frequency and electro-optic methods. RoF emerges as a promising solution to fulfill the capacity and mobility of access networks providing high data rate, high capacity and a mobility proposal to transmit information [2-4]. Given that the transmission of information is via light through optical fibers, it is evident that its inherent advantages are exploited, such as low losses, immunity to electromagnetic interference and especially the huge bandwidth available. Figure 1 shows a basic RoF scheme composed by a RF-to-Optical stage, an optical link and an Opticalto-RF stage. The RF signal modulates the light (either directly or externally). The modulated light travels across the optical channel. Finally, the light is turned into its electrical form where it can be radiated by an antenna providing wireless coverage in zones of difficult connection such as shopping centers and subterraneous places [5].

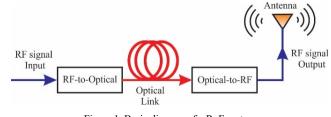


Figure 1. Basic diagram of a RoF system

RoF technology is a candidate that aims to alleviate the bandwidth restrictions for 5G Mobile, delivering wireless data signals via optical fiber channels. [6]. Thus, RoF allows supporting secure, cost-effective coverage as well as excellent capacity mobile-wireless access for oncoming services. Some drawbacks inherent to RoF are due to the distortion and noise derived by the non-linear features of the components employed in the optical network as the optical sources, the chromatic dispersion and photodetectors. For this reason, to mitigate sources of noise and

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misstatement, intensity-modulation-direct-detection is advisable [7]. In this regard, we propose and experimentally demonstrate a RoF scheme using an optical communication system in the modality of external modulation for the transmission and distribution of a RF-signal. Furthermore, it is demonstrated that the chromatic dispersion (considered as a drawback in this type of systems), is used in an advantageous manner to the generation of notch filters. On the other side, it is known that an XBee is an electronic device whose operation is based on the network transport protocol IEEE 802.15.4, it is able to generate wireless paths to establish communication between devices (http://xbee.cl/que-es-xbee, https://www.digi.com/products/xbeerf-solutions/xctu-software/xctu). In this sense, XBee modules that use the ZigBee protocol find applications in RoF networks [8, 9]. In this experimental proposal, a wireless signal delivered by an XBee module is used as test signal. A baseband modulation Gaussian Frequency Shift Keying (GFSK) signal at 2.44GHZ is generated by an XBee. This electrical signal modulates an intensity electro-optic modulator. This work is structured as follows. In Section 2, is given the principle of operation of the optical communication system at external modulation and also numerical simulations are presented. In Section 3, experimental set-up of the RoF is described. At the end, a short conclusion is given in Section 4.

2. Principle of Operation

Figure 2 depicts the block diagram the optical communication system in the modality of external modulation used in this work. Basically, it is constituted by a single-mode laser diode as optical source, a Mach Zhender-Intensity Modulator (MZ-IM), a Single Mode-Standard Fiber (SM-SF), as well as a Photodiode (PD).

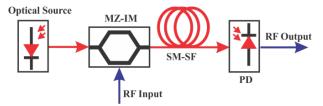


Figure 2. Block diagram of the model used

In the following, a full mathematical analysis explaining the behavior of the model is given. In order to exploit the chromatic dispersion exhibited by the SM-SF, it is necessary to assume that the single-mode laser diode emits at 1550nm. It is demonstrated that the chromatic dispersion (D) and the length (L) of the optical link play an important role that allow the system to act as photonic notch filter.

Mathematically, the optical spectrum of a single-mode laser diode (for example a DFB) can be modeled by a Gaussian envelope centered at an angular frequency ω_0 as [10]

$$S_0(\omega - \omega_0) = \frac{2S_{max}}{\Delta_\omega \sqrt{\pi}} exp\left(\frac{-4(\omega - \omega_0)^2}{\Delta_\omega^2}\right) \tag{1}$$

where S_{max} is the maximum power emission and Δ_{ω} is the Full Width at Half Maximum of the spectrum that can be expressed in terms of $\Delta\lambda$. Eq. (1) can be simplified, defining $a = 4/\Delta_{\omega}^2$ and $W = \omega - \omega_0$, resulting in

$$S_0(W) = \sqrt{\frac{a}{\pi}} S_{max} exp(-aW^2)$$
(2)

The Fourier Transform (FT) of Eq. (2) is the spectral density of the optical source, thus

$$\{S_0(W)\} = \int_0^\infty S_0(W) exp(-j2\pi ZW) dW \qquad (3)$$

Defining $Z = v_m \beta_2 L$, and expressing the propagation constant $\beta_2 = -D \frac{\lambda^2}{2\pi c}$, defining $\omega_m = 2\pi v_m$, and $dW = d\omega$, then

$$\mathcal{F}\{S_0(W)\} = \int_{-\infty}^{\infty} \sqrt{\frac{a}{\pi}} S_{max} exp(-aW^2) exp(-j2\pi ZW) dW$$
⁽⁴⁾

Equation (4) is solved considering that the *FT* of a Gaussian function $f(t) = exp(-at^2)$ is [11]

$$\mathcal{F}[exp(-at^2)] = \sqrt{\frac{\pi}{a}} exp\left(-\frac{\omega^2}{4a}\right) \tag{5}$$

Rewriting Eq. (4) and considering a, and Z we find

$$\mathcal{F}\{S_0(W)\} = S_{max} exp\left(-\frac{(\pi \nu_m \beta_2 L \Delta_\omega)^2}{4}\right) \tag{6}$$

When $S_0(W)$ is narrow, the frequency response of the system is

$$H(\nu_m) = \cos\left(\frac{\pi\nu_m^2\lambda^2 DL}{c}\right) \tag{7}$$

The squared of the modulation frequency v_m^2 makes the system behave like a non-periodic notch filter.

The notch frequencies f_i , for which Eq. (7) becomes zero, are determined by

$$f_i = \frac{1}{\lambda} \sqrt{\frac{1}{2} \frac{(2i-1)c}{DL}}$$
(8)

where λ is the central wavelength in nm, *D* is in ps/nm·km, *c* is speed of light in nm/s, and *L* is the length of the optical link in km.

Equation (7) is numerically evaluated by using MATLAB for the frequency range 0-25GHz, considering: λ =1550nm, c=2.9979x10¹⁷nm/seg, L=25.24km, D=15.81ps/nm·km.

Moreover, substituting λ , c, L, and D in Eq. (8), the first and second notch frequency values are $f_1=12.50$ GHz and $f_2=21.65$ GHz, respectively.

Figure 3 shows the result of this numerical evaluation. The result of this evaluation is corroborated by a block level simulation using VPIphotonics software® (http://www.vpiphotonics.com/index.php), and plotted on the same graph. This software is provided with a full library of electro-optics components whose properties can be easily manipulated, therefore, the degree of reliability for the results is high.

From this graph, it is clearly appreciable that the separation between consecutive notches is not constant and decreases as the frequency increases. The bandwidth at -3dB for the low-pass band and for the first lobe (band-pass) are in the order of GHz. The common point between both curves is their tendency. Note that both curves converge to the values of 12.50GHz and 21.65GHz as stablished by Eq. (8). The justification to use 25km is because the current transmission lengths are designed to meet 20 km, as is determined by an international standard [12].

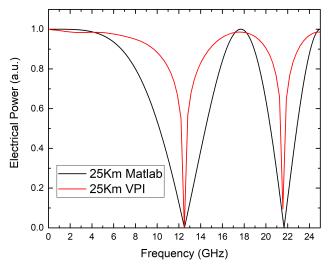


Figure 3. Simulated frequency response using MATLAB and VPI photonics software.

3. Experimental Results

Initially, a test communication between two XBee devices is carried out. Later, a characterization of the optical communication system is done. Finally, the performance of the system composed by a couple of XBee modules over a RoF link is accomplished.

3.1 Test of the XBee devices

XBee is a trademark of radio devices capable of supporting communication protocols as ZigBee, 802.15.4, and Wi-Fi, among others [9]. Figure 4 depicts an XBee network. One module plays the role of Coordinator whereas the other of an End Device.

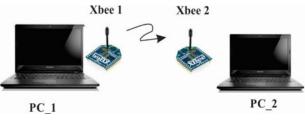


Figure 4. Network generated by using two XBee modules.

The Coordinator handles out addresses and manages the network The End Device is responsible for joining existing networks, transmitting and receiving information. In this work, two XBee modules (XBee-PRO model ZNet 2.5 OEM, indoor range 100m, operating frequency 2.4GHz, transmit power output 18dBm) are used. Both devices are provided with an omnidirectional antenna (Gain 2.15dBi, frequency range from 2.4 to 2.4835GHz) and an USB adapter (Sparkfun XBee Explorer Dongle) that allows the connection via serial communication. The communication between these devices is achieved by using the platform XCTU. Figure 5 shows this interface (https://www.digi.com/products/xbee-rf-solutions/xctu-

software/xctu). Data to be transmitted by PC_1 (Coordinator) are coded at the frequency of 2.44GHz and then radiated via an antenna and decoded by PC_2 (End Device).

Close Record Detach		CTS CD DSR DTR RTS BRK										Rx Bytes: 0													
Console log																					0	0	0	0	C
This a Test messageThis a	54	68	69	73	20	61	20	54	65	73	74	20	6D	65	73	73	61	67	65	54	68	69	73	20	
Test messageThis a Test	61	20	54	65	73	74	20	6D	65	73	73	61	67	65	54	68	69	73	20	61	20	54	65	73	
messageThis a Test	74	20	6D	65	73	73	61	67	65	54	68	69	73	20	61	20	54	65	73	74	20	6D	65	73	
messageThis a Test	73	61	67	65	54	68	69	73	20	61	20	54	65	73	74	20	6D	65	73	73	61	67	65	54	
messageThis a Test	68	69	73	20	61	20	54	65	73	74	20	6D	65	73	73	61	67	65	54	68	69	73	20	61	
messageThis a Test	20	54	65	73	74	20	6D	65	73	73	61	67	65	54	68	69	73	20	61	20	54	65	73	74	
messageThis a Test message	20	6D	65	73	73	61	67	65																	

Figure 5. Screenshot showing the interface to control the Coordinator.

Figure 6 shows the spectrum of the emitted radio frequency signal at 2.44GHz measured by an Electrical Signal Analyzer (ESA) exhibiting a SNR value of 55dB. The baseband data is coded within the side lobes.

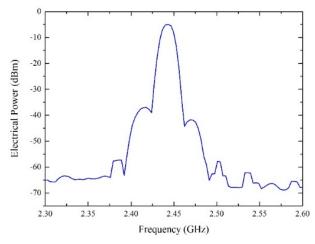


Figure 6. Electrical spectrum corresponding to the Coordinator

3.2 Experimental Frequency Response

Figure 7 illustrates the experimental bench used in this work. It is well known that the use of multimode optical fibers is reserved for short distances, such as in local area networks. Because, we are interested to emulate a FTTH-PON, whose typical length is in the order of 20km [12], thus the use of single-mode optical fiber is mandatory, due to its characteristics of extremely low-losses. Furthermore, most of commercial electro-optical devices (optical sources, optical modulators, photodetectors, etc.) operating at the commercial wavelengths of 1300 and 1550 nm, are supplied with this type of fiber. From the comments above described, SM-SF is used in this experiment. In the next sub-section, a detailed description of the optical communication system is given.

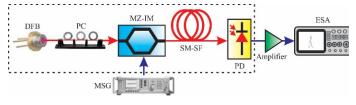


Figure 7. Optical communication system externally modulated

First, the optical spectrum of a DFB (NX8508-55, Side Mode Suppression Ratio SMSR=40dB, linewidth 0.1nm) is recorded by means of an Optical Spectrum Analyzer. According to the datasheet, this optical source has its own internal optical isolator to avoid reflections, therefore a good optical stability is guaranteed. Figure 8 shows the optical spectrum for an operation current of

20mA exhibiting a central wavelength λ_0 =1553.5nm. The shape of the optical spectrum corresponds to Eq. (1).

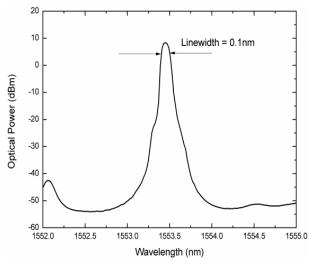


Figure 8. Optical spectrum of the DFB laser source used

Once the characteristics of the optical source are known, the light generated by the DFB is modulated via the electro-optic MZ-IM (insertion loss of 2.7dB, BW-20GHz, operating wavelength 1530-1580 nm, $V\pi$ =7.5 V). The Polarization Controller (PC) device allows optimizing the optical power level at the output of the modulator. The RF signal used to modulate the light is provided by a Microwave Signal Generator (MSG) through a sweep in frequency range of 0.01-13 GHz at a power of 0 dBm. The modulated beam is injected to a reel of SM-SF (D=15.81ps/nm·km (a) 1550nm, α =0.22dB/km, L=25.24 km). The light that emerges of the optical fiber is turned to its corresponding electrical signal by the PD (Miteq, BW-13GHz, Responsivity=0.9Amp/Watt). This signal is submitted to a process of amplification (Minicircuits, ZVA-183-S+ Ultra-Wideband Amplifier, 0.7-18GHz, Gain 26dB) and finally visualized in the screen of the ESA. Figure 9 corresponds to the measured frequency response where the simulation curves have been superimposed. Since the frequency emitted by the XBee Module is at 2.44GHz, the low-pass band will be used to code the transmission of the RF signal.

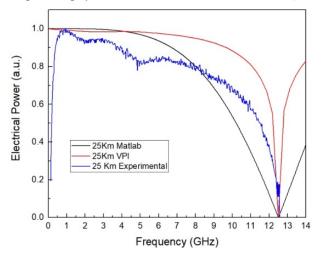


Figure 9. Comparison between simulated and experimental results

From this graph, it is noticeable that the response is restricted by the BW of the PD.

3.3 Experimental RoF Transmission

Figure 10 shows the test bench for carry out the transmission of the signal issued from the XBee module. The main difference to the scheme shown in Fig. 7 is that now the MSG is replaced by the coordinator module connected to PC 1 generating the information to be sent (an electrical signal at 2.44GHz at -6dBm) that is emitted from Antenna 1. Antenna 2 receives this signal which is amplified and launched to the RF port of the MZ-IM to modulate the optical signal. The modulated light is injected to the SM-SF. At the end of the link, the modulated light is converted into electric signal by the PD. Subsequently, amplified and separated by a power divider. A percentage of the recovered signal is evaluated by the ESA and the remainder is connected to the Antenna 3 where is emitted. An anechoic chamber is used to guarantee isolation between the signals emitted by antennas 1 and 3. Lastly, antenna 4 captures the signal to be handled by the second XBee module coupled to computer number 2 (PC 2), recovering in this manner the original transmitted message. Figure 11 corresponds to the measured electrical spectrum corresponding to the recovered signal with a SNR of 27dB.

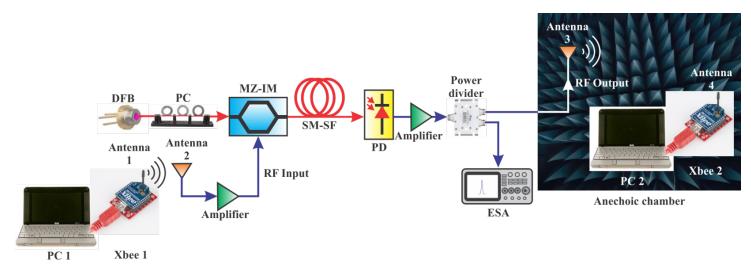


Figure 10. Experimental scheme assembled to test the RoF XBee transmission system

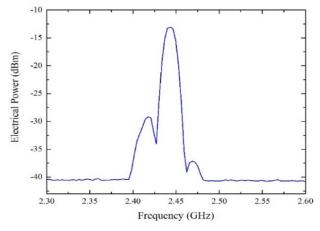


Figure 11. Recovered spectrum corresponding to the output of the system

4. Conclusion

In this work, we have experimentally demonstrated an efficient communication between two XBee modules using the RoF technique. The optical transmission was carried out by using an external modulation scheme. A full mathematical analysis explaining the principle of operation of an optical communication system was given. The relationship that allows determining the frequency response of the optical communication system was evaluated by a numerical simulation achieved in Matlab. Later, a block level simulation of the optical system was carried out by using VPIphotonics software corroborating the previous results. One thing to keep in mind is that the usage of an optical source whose central wavelength is 1310nm associated with SMSF allows the cancellation of the chromatic dispersion effect [12]. However, in this work, we have used a DFB laser whose central wavelength is around 1550nm and SM-SF; thus, the chromatic dispersion value was considerable. This effect, as well as, the length of the optical link has an important role to determine the electrical bandwidth of the system as it was evident in Eq. (7) and Eq. (8). The frequency response was composed by a low-pass band as well as a series of bandpass or notches. In particular, the first lobe was used for the transmission of a wireless signal of 2.44GHz issued by the XBee module. Optical transmission was successfully achieved through an optical link of 25.24Km. The recovered RF signal exhibits a SNR of 27dB. Considering that the electrical bandwidth can be tailored in function of the length of the optical link, it is possible to assure the possibility to accommodate future services distributed by means of RoF schemes using a larger bandwidth. Furthermore, another interesting feature of this RoF system is that, at the end of the link, data can be distributed by antennas to several customers. Direct and external modulation can be used to implement a D-RoF system in our case we have opted for external modulation using a Mach Zehnder modulator (MZM) to modulate a DFB laser in order to generate a simpler configuration [13]. As future work, we propose the use of a VCSEL as optical source in order to benefit of its properties as threshold currents of very few mA and facilitate the coupling with optical fibers [14].

Acknowledgments

L. A. González-Mondragón, L. J. Quintero-Rodríguez, and A. G. Correa-Mena, wish to thank the CONACyT for the student scholarships number 553921, 465594 and 335148, respectively. www.astesj.com

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